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ABSTRACT

The experimental characterization of the discontinuities arising at the junction between two uniform microstrips having different characteristic impedances is obtained through a non-resonant method able to give broadband information.

In the present work we expose a non resonant method able to give broadband information on microstrip discontinuities. A schematic view of the structure used in the experiments is shown in Fig.1: a tract of uniform microstrip of characteristic impedance Z_1 , l_1 in length, is attached to two tracts of lengths l_0 and characteristic impedance Z_0 (by "characteristic impedance" we mean the quasi-static value).

The composed microstrip structure is connected to the measuring instrument through two launchers L_1 , L_2 :

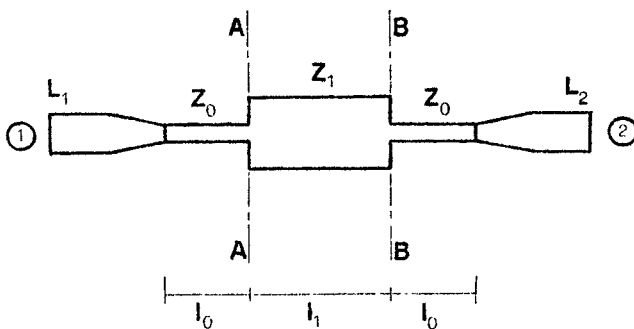


Fig. 1

In our actual experiments the measuring instrument was an 8542A Automatic Network Analyzer (Hewlett-Packard). The discontinuities at A-A, B-B are to be characterized, and this is done by a de-embedding procedure summarized in the following.

Firstly one should know the scattering parameters of both launchers. This is possible by a method recently developed¹: one measures at the coaxial ports ① and ② of the launchers the two systems composed of the cascade launcher L_1 - a tract $l(k)$ ($k=1,2$) of microstrip of characteristic impedance Z_0 - launcher L_2 ; in the quoted paper¹ it is shown that, from measurements taken on such systems, by a suitable mathematical treatment one can calculate at any frequency of interest (a) the propagation constant of the microstrip, (b) the full scattering matrix of L_1 , including the launcher-microstrip discontinuity; and similarly for L_2 . It is shown¹ that the variation of the characteristic impedance with the frequency can be neglected, in the sense that its effects can be included in the launchers themselves.

Once L_1 , L_2 and the two tracts of length l_0 are characterized, the de-embedding procedure on the structure of Fig.1 works as follows:

- the scattering matrix S^* at ports ① and ② is measured
 - from S^* the effects of L_1 , L_2 and of the tracts of lengths l_0 are subtracted and finally one obtains the scattering matrix S of the structure composed of: (a) the discontinuity in A-A, (b) the uniform tract of impedance Z_1 and (c) the discontinuity in B-B.
- Fig. 2 shows a result obtained by the above procedure. It refers to a case with $Z_0 = 50$ ohm, $Z_1 = 14.88$ ohm and a dielectric constant of the substrate (Alumina) 9.7. The figure shows, in the 2-10 GHz band, the plots of $|S_{11}^*|$ (dotted line) and of $|S_{11}|$ (full line). It is apparent from this example that the launcher effects are by no means negligible.

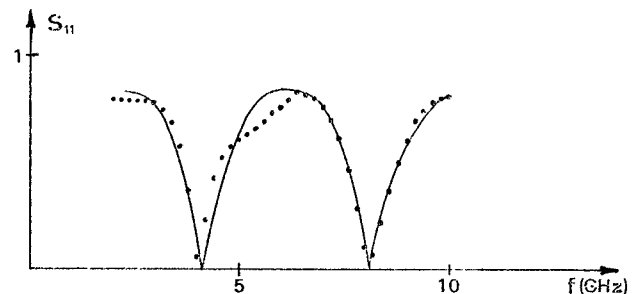


Fig. 2

Finally, in order to characterize the discontinuities, a standard lossless T network is assumed to represent them. The discontinuity in A-A is modelled by a shunt capacitor C inserted between two series inductors L_a , L_b , whereas for B-B the same circuit is considered, with the inductors in reverse order.

Now, since matrix S is known, the calculation of L_a , L_b and C is, in principle, a simple matter. In fact the tract of uniform microstrip between A-A and B-B has a calculable scattering matrix: the dispersion is taken into account by using a model due to Carlin² which was found highly accurate³. Hence the three unknown parameters can be calculated by imposing that the overall matrix of the cascade "discontinuity A-A - uniform tract l_1 - discontinuity B-B" be identical to the matrix S found through the de-embedding procedure. To minimize errors,

the three parameters were actually determined by an optimization procedure based on Fletcher's algorithm⁴. For $Z_0 \approx 50$ ohm, $Z_1 = 14.88$ ohm, $\epsilon_r = 9.7$ resulted

$$L_a = 0.027 \text{ nH}$$

$$L_b = 0.127 \text{ nH}$$

$$C = 0.115 \text{ pF}$$

The above values are in reasonable agreement with other found in the literature. The characterization of discontinuities arising with other impedance values is in progress.

References

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